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TITLE: FIRST AND SECOND DERIVATIVE PROCESSING OF
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**FIRST AND SECOND DERIVATIVE PROCESSING
OF WAVELENGTH MULTIPLEXED OPTICAL SIGNALS**

BACKGROUND OF THE INVENTION:

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Field of the Invention

The present invention relates generally to optical devices such as lasers, and fiber optic data transmission systems employing the same, and particularly to a novel wavelength-
locked loop servo-control circuit for optimizing performance of optical signal processing equipment, especially equipment employed in wavelength division multiplexing (WDM) systems and dense wavelength division multiplexing (DWDM) systems.

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Description of the Prior Art

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Wavelength Division Multiplexing (WDM) and Dense Wavelength Division Multiplexing (DWDM) are light-wave application technologies that enable multiple wavelengths (colors of light) to be paralleled into the same optical fiber with each wavelength potentially assigned its own data diagnostics. Currently, WDM and DWDM products combine many different data links over a single pair of optical fibers by re-modulating the data onto a set of lasers, which are tuned to a very specific wavelength (within 0.8 nm tolerance, following industry standards). On current products, up to 32 wavelengths of light can be combined over a single fiber link with more wavelengths contemplated for future applications. The wavelengths are combined by passing light through a series of thin film interference filters, which consist of multi-layer coatings on a glass substrate, pigtailed with optical fibers. The filters combine multiple wavelengths into a single fiber path, and also separate them again at the far end of the multiplexed link. Filters may also be used at intermediate points to add or drop wavelength channels from the optical network.

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A key factor in determining the ultimate detector sensitivity and bit error rate in fiber optic wavelength multiplexing systems is spectral interference at the wavelength of interest caused by overlap between adjacent wavelength bands. This form of optical crosstalk is especially important in DWDM systems, where the wavelength spacing is
5 currently standardized at 0.8 nm and may be reduced to as little as 0.4 nm or less on next generation systems. In addition to applications in the design of WDM networking equipment, this approach is very valuable in the design of WDM optical test equipment. Weak signals at the wavelength of interest may also be nested in broadband background noise, which also limits their detection by decreasing ambient signal-to-noise ratio. This
10 problem is additionally pertinent to spectroscopy and other forms of optical signal processing equipment.

It would thus be highly desirable to provide a system and method for automatically improving detection sensitivity in WDM and DWDM systems.

15 One technique which may potentially improve the detection limit by more than an order of magnitude involves measuring either the first or second derivative of the optical transmission curve with respect to wavelength. The derivative output signal is directly proportional to the optical loss (for example, due to absorption, impurities, or crosstalk)
20 in the wavelength communication channel.

It would thus be highly desirable to provide an apparatus and method for implementing derivative measurement techniques in a practical apparatus compatible with existing WDM and DWDM network equipment.

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SUMMARY OF THE INVENTION

It is therefore an object of the present invention to provide a system and method for automatically improving signal detection sensitivity in WDM and DWDM systems.

It is another object of the present invention to provide a system and method for implementing derivative measurement techniques in a practical apparatus compatible with existing WDM and DWDM network equipment.

- 5 It is a further object of the present invention to provide a system and method for implementing derivative measurement techniques for increasing signal detection sensitivity in WDM and DWDM systems that employs a servo-control loop that enables dynamic tracking of the center wavelength of an optical signal to the peak of an optical filter passband.
- 10 It is still another object of the present invention to provide a servo/feedback loop, referred to as a "wavelength-locked loop," for use in implementing derivative measurement techniques in a practical apparatus compatible with existing WDM and DWDM network equipment.
- 15 Thus, according to one aspect of the invention, there is provided a system and method for detecting optical signals in optical networks comprising: an optical signal generator for generating optical signals, each optical signal having a peaked spectrum function including a center wavelength for transmission over a communication channel; a
- 20 mechanism for applying a dither modulation signal at a dither modulation frequency to the optical signal about the center wavelength; and, a wavelength-locked loop servo-control circuit for detecting a rate of change of an intensity of the dither modulated optical signal with respect to the center wavelength, the detected rate of change indicating a degree of optical attenuation in a communication system at that wavelength. The
- 25 wavelength-locked loop servo-control circuit further enables real-time adjustment of the optical signal center wavelength in a manner so as to minimize optical signal attenuation in the communication channel.

According to a further aspect of the invention, there is provided a system and method for detecting optical signals in optical networks, the optical network including a receiver portion for receiving optical signals transmitted over a communications channel, the system comprising: an optical signal generator for generating optical signals, each optical
5 signal having a peaked spectrum function including a center wavelength; a tunable wavelength selective device for receiving optical signals transmitted over the communication channel, the wavelength selective device nominally exhibiting a peaked passband function including a center wavelength; a mechanism for applying a dither modulation signal to the wavelength selective device for dithering the peaked passband
10 function of the tunable wavelength selective device about the center wavelength, the dither modulated tunable wavelength selective device causing generation of a modulated optical signal at the dither modulation frequency; and, a wavelength-locked loop servo-control circuit for detecting a rate of change of an intensity of the dither modulated optical signal with respect to the center wavelength, the detected rate of change indicating
15 a degree of optical attenuation in the communication channel at that wavelength. The wavelength-locked loop servo-control circuit enables real-time center wavelength adjustment of the peaked passband function of the tunable wavelength selective device in a manner so as to minimize a degree of optical signal attenuation in the communication channel.

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BRIEF DESCRIPTION OF THE DRAWINGS

Further features, aspects and advantages of the apparatus and methods of the present invention will become better understood with regard to the following description,
25 appended claims, and the accompanying drawings where:

Figure 1(a) depicts a system block diagram depicting the underlying system architecture employing the wavelength-locked loop technique in an optical system providing first and

second derivative processing of wavelength multiplexed optical signals according to a first embodiment of the present invention;

5 Figure 1(b) depicts a system block diagram depicting the underlying system architecture employing the wavelength-locked loop technique in an optical system providing first and second derivative processing of wavelength multiplexed optical signals according to a second embodiment of the present invention;

10 Figures 2(a) and 2(b) depict example underlying wavelength-locked loop system architectures;

15 Figure 2(c) is a general block diagram depicting the underlying system architecture for tuning tunable frequency selective devices such as a bandpass filter according to the principles of the present invention;

Figures 3(a)–3(b) are signal waveform diagrams depicting the relationship between laser optical power as a function of wavelength for three instances of optic laser signals;

20 Figures 4(a)–4(c) are signal waveform diagrams depicting the laser diode drive voltage dither modulation (a sinusoid) for each of the three waveform diagrams of Figures 3(a)–3(c);

25 Figures 5(a)–5(c) are signal waveform diagrams depicting the resulting feedback error signal output of the PIN diode for each of the three waveform diagrams of Figures 3(a)–3(c);

Figures 6(a)–6(c) are signal waveform diagrams depicting the cross product signal resulting from the mixing of the amplified feedback error with the original dither sinusoid; and,

Figures 7(a) - 7(c) are signal waveform diagrams depicting the rectified output laser bias voltage signals which are fed back to adjust the laser current and center frequency.

DETAILED DESCRIPTION OF PREFERRED EMBODIMENTS

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The present invention is directed to techniques for measuring the optical loss of a communication channel due to absorption, impurities, or crosstalk, for example, at specified wavelength by determining either the first or second derivative of the fiber or channel's optical transmission curve with respect to the wavelength. This is due to the 10 recognition that the derivative output signal is directly proportional to the optical loss in the wavelength communication channel. This invention is particularly directed to a system and method for implementing derivative measurement techniques by employment of a novel servo-control system in practical optical signal processing equipment, e.g., equipment compatible with existing WDM and DWDM network equipment.

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The explanations herein discuss both wavelength and frequency, which have a reciprocal relationship ($\lambda = c/f$, where c = speed of light), as is well known in the field of optics.

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As shown in Figure 2(a), the novel servo-control system implements a principle referred to herein as the "wavelength-locked loop" or "lambda-locked loop" (since the symbol lambda is commonly used to denote wavelength). The basic operating principle of the wavelength-locked loop (WLL) is described in greater detail in commonly-owned, co-pending U.S. Patent Application Serial No. 09/865,256, entitled APPARATUS AND METHOD FOR WAVELENGTH-LOCKED LOOPS FOR SYSTEMS AND 25 APPLICATIONS EMPLOYING ELECTROMAGNETIC SIGNALS, the whole contents and disclosure of which is incorporated by reference as if fully set forth herein.

Particularly, as described in commonly-owned, co-pending U.S. Patent Application Serial No. 09/865,256, and with reference to Figure 2(a), the wavelength-locked loop principle

implements a dither modulation to continuously adjust an electromagnetic signal source characterized as having a peaked frequency spectrum or peaked center wavelength, e.g., a laser light source, so as to track the center of a frequency selective device, e.g. a filter passband. In this manner, optimal power of the signal is transmitted and optimal use is

- 5 made of the system transmission bandwidth. The principle may be exploited for tuning any light source having a peaked frequency spectrum, and additionally, may be used to tune or adjust transmission properties of frequency selective devices such as tunable filters.

- 10 For purposes of description, the basic operating principle of the WLL is shown in Figure 2(a) which depicts an example optic system 10 including a light source such as laser diode 12 driven with both a bias voltage 15 from a voltage bias circuit 14, and modulated data 18 from a data source (not shown). The laser diode generates an optical (laser light) signal 20 that is received by a bandpass filter 25 or, any frequency selective device 15 including but not limited to: thin film optical interference filters, acousto-optic filters, electro-optic filters, diffraction gratings, prisms, fiber Bragg gratings, integrated optics interferometers, electroabsorption filters, and liquid crystals. The laser diode itself may comprise a standard Fabry Perot or any other type (e.g., Vertical Cavity Surface Emitting (VCSEL)), light emitting diodes, or, may comprise a Distributed Feedback 20 semiconductor laser diode (DFB) such as commonly used for wavelength multiplexing. Preferably, the laser diode emits light in the range of 850 nm to 1550 nm wavelength range. As mentioned, the bandpass filter may comprise a thin film interference filter comprising multiple layers of alternating refractive indices on a transparent substrate, e.g., glass. As further shown in Figure 2(a), according to the invention, there is an added 25 sinusoidal dither modulation circuit or oscillator 22 for generating a sinusoidal dither modulation signal 27 that modulates the laser bias voltage. The sinusoidal dither signal may be electronically produced, e.g., by varying the current for a laser, or mechanically, by varying the micro-electromechanical system's (MEMS) mirror to vary the wavelength. The dither modulation frequency is on the order of a few kilohertz (kHz) but

may range to the Megahertz range. Preferably, the dither modulation frequency is much less than the data rate which is typically on the order of 1-10 GHz. Modulation of the laser diode bias current 15 in this manner causes a corresponding dither in the laser center wavelength. Modulated data is then imposed on the laser, and the optical output passes

5 through the bandpass filter 25. Preferably, the filter 25 is designed to tap off a small amount of light 29, for example, which is incident upon a photo detector receiver device, e.g., P-I-N diode 30, and converted into an electrical feedback signal 32. The amount of light that may be tapped off may range anywhere between one percent (1%) to five

10 percent (5%) of the optical output signal, for example, however, skilled artisans will appreciate any amount of laser light above the noise level that retains the integrity of the output signal including the dither modulation characteristic, may be tapped off. The remaining laser light passes on through the filter 25 to the optical network (not shown).

As the PIN diode output 32 is a relatively weak electric signal, the resultant feedback signal is amplified by amplifier device 35 to boost the signal strength. The amplified

15 electric feedback signal 37 is input to a multiplier device 40 where it is combined with
d.s. the original dither modulation signal ³⁵₃₅. The cross product signal 42 that results from the
d.s. multiplication of the amplified PIN diode output (feedback signal) 37 and the dither
signal ³⁵₃₅ includes terms at the sum and difference of the dither frequencies. The result is

thus input to a low pass filter device 45 where it is low pass filtered and then averaged by

20 integrator circuit 48 to produce an error signal 50 which is positive or negative depending on whether the laser center wavelength is respectively less than or greater than the center point of the bandpass filter. The error signal 50 is input to the laser bias voltage device 15 where it may be added (e.g., by an adder device, not shown) in order to correct the laser bias current 15 in the appropriate direction. In this manner, the bias current (and

25 laser wavelength) will increase or decrease until it exactly matches the center of the filter passband. Alternately, the error signal 50 may be first converted to a digital form, prior to input to the bias voltage device.

According to one aspect of the invention, the WLL will automatically maintain tracking of the laser center wavelength to the peak of the optical filter. However, in some cases, it may not be desirable to enable laser alignment to the filter peak, e.g., in an optical attenuator. Thus, as shown in Figure 2(b) which is a system 10' corresponding to the system 10 of Figure 2(a), there is provided an optional external tuning circuit, herein referred to as a wavelength shifter device 51, that receives the error signal and varies or offsets it so that the laser center wavelength may be shifted or offset in a predetermined manner according to a particular network application. That is, the wavelength shifter 51 allows some external input, e.g., a manual control element such as a knob, to introduce an arbitrary, fixed offset between the laser center wavelength and the filter peak.

It should be understood that, as described in commonly-owned, co-pending U.S. Patent Application Serial No. 09/865,256, the WLL servo-control system may be implemented for tuning tunable frequency selective devices such as a bandpass filter for a variety of optical network applications. Thus, in the embodiment depicted in Figure 2(c), the system 10" comprises similar elements as system 10 (of Figure 2(a)) including a bias voltage generator device 14 for applying a bias signal 15 to the laser diode 12 for generating an optical signal 20 having a peaked spectrum function. This signal 20 is input to a tunable frequency selective device 25, e.g., a tunable bandpass filter. As shown in Figure 2(c), however, the sinusoidal dither/driver device 22 is implemented for modulating the peak center frequency of filter pass band with a small dither signal 27. A small amount of light 29 is tapped off the output of the filter 25 for input to the photodetector device, e.g., PIN diode 30, where the optical signal is converted to electrical signal 32, amplified by amplifier device 35, and input to the mixer device 40 which additionally receives the dither signal 27. The mixer device generates the vector cross product 42 of the amplified feedback signal 37 with the dither signal 27 and that result is low-pass filtered, and smoothed (e.g., integrated) by integrator device 48 to provide error signal 50, in the manner as will be discussed herein with reference to Figures 3-7. This error signal 50 may be a bi-polar signal and may be used to

dynamically adjust the peak center frequency of the filter passband until it matches the center frequency of the laser signal input 20.

The operating principle of the WLL is further illustrated in the timing and signal 5 diagrams of Figures 3–7. Figures 3(a)–3(c) particularly depicts the relationship between laser optical power as a function of wavelength for three instances of optic laser signals: a first instance (Figure 3(a)) where the laser signal frequency center point 21 is less than the bandpass function centerpoint as indicated by the filter bandpass function 60 having centerpoint 62 as shown superimposed in the figures; a second instance (Figure 3(b)) 10 where the laser frequency center point 21 is aligned with the bandpass function centerpoint 62; and, a third instance (Figure 3(c)) where the laser frequency center point 21 is greater than the bandpass function centerpoint 62. In each instance, as depicted in corresponding Figures 4(a)-4(c), the laser diode drive voltage signal 15 is shown dithered (a sinusoid) resulting in the laser wavelength dithering in the same manner. The dithered 15 laser diode spectra passes through the filter, and is converted to electrical form by the PIN diode 30. In each instance of the laser signals depicted in Figures 3(a) and 3(c) having frequency centerpoints respectively less than and greater than the band pass filter centerpoint, it is the case that the dither harmonic spectra does not pass through the frequency peak or centerpoint of the bandpass filter. Consequently, the resulting output 20 of the PIN diode is an electric sinusoidal signal of the same frequency as the dither frequency such as depicted in corresponding Figures 5(a) and 5(c). It is noted that for the laser signals at frequencies below the peak (Figure 3(a)) the feedback error signal 32 corresponds in frequency and phase to the dither signal (Figure 5(a)), however for the laser signals at frequencies above the peak (Figure 3(c)) the feedback error signal 32 25 corresponds in frequency but is 180° opposite phase of the dither signal (Figure 5(c)). Due to the bipolar nature of the feedback signal (error signal) for cases when the laser signal centerpoint is misaligned with the bandpass filter centerpoint, it is thus known in what direction to drive the laser diode (magnitude and direction), which phenomena may be exploited in many different applications. For the laser signal depicted in Figure 3(b)

having the laser frequency center point aligned with the bandpass function centerpoint, the dither harmonic spectra is aligned with and passes through the frequency peak (maximum) of the bandpass filter twice. That is, during one cycle (a complete round trip of the sinusoid dither signal), the dither signal passes though the centerpoint twice. This
5 results in a frequency doubling of the dither frequency of the feedback signal 32, i.e., a unique frequency doubling signature, as depicted as PIN diode output 32' in Figure 5(b) showing an feedback error signal at twice the frequency of the dither frequency.

Thus, in each instance, as depicted in corresponding Figure 5(b), the resulting feedback
10 signal exhibits frequency doubling if the laser center wavelength is aligned with the filter center wavelength; otherwise it generates a signal with the same dither frequency, which is either in phase (Figure 5(a)) or out of phase (Figure 5(c)) with the original dither modulation. It should be understood that, for the case where the laser center frequency is misaligned with the bandpass filter peak and yet there is exhibited partial
15 overlap of the dither spectra through the bandpass filter peak (i.e., the centerpoint peak is traversed twice in a dither cycle), the PIN diode will detect partial frequency doubling laser at opposite phases depending upon whether the laser center frequency is inboard or outboard of the filter center frequency. Thus, even though partial frequency doubling is detected, it may still be detected from the feedback signal in which direction and
20 magnitude the laser signal should be driven for alignment.

Referring now to Figures 6(a) and 6(c), for the case when the laser and filter are not aligned, the cross product signal 42 resulting from the mixing of the amplified feedback error with the original dither sinusoid is a signed error signal either at a first polarity (for
25 the laser signals at frequencies below the bandpass filter centerpoint), such as shown in Figure 6(a) or, at a second polarity (for the laser signals at frequencies above the bandpass filter centerpoint), such as shown in Figure 6(c). Each of these signals may be rectified and converted into a digital output laser bias voltage signal 48 as shown in respective Figures 7(a) and 7(c), which are fed back to respectively increase or decrease

the laser current (wavelength) in such a way that the laser center wavelength moves closer to the bandpass filter centerpoint. For the case when the laser and filter are aligned, the cross product generated is the frequency doubled signal (twice the frequency of the dither) as shown in the figures. Consequently, this results in a 0 V dc bias voltage
5 (Figure 7(b)) which will maintain the laser frequency centerpoint at its current wavelength value.

The system for measuring the optical loss of a communication channel, e.g., due to absorption, impurities, or crosstalk, at specified wavelength by determining either the
10 first or second derivative of the fiber or channel's optical transmission curve with respect to the wavelength according to the present invention is now described. Principally, a measurement is made by rapid repetitive scanning, or dithering, of a narrow wavelength interval centered on an optical signal wavelength of interest. This periodic scan is synchronized with a tuned A.C. amplifier locked to the fundamental or harmonic multiple
15 of a dither frequency. The modulated signal which results is the first derivative of optical intensity with respect to wavelength when the amplifier is locked to the dither frequency. At twice this frequency, the second derivative may be obtained. The signal-to-noise and signal-to-background ratios are increased by bandpass filter rejection of the low-frequency "flicker noise", commonly called $1/f$ noise, and the power line interference
20 noise associated with electronic components in the system, rectifiers on the card, or some types of optical sources. Thus, noise sources of different frequencies than the modulation frequency will not pass through the phase-sensitive lock-in amplifier. Similarly, by amplitude modulation of the light beam intensity or "chopping" the source, the D.C. offset and drift (for example, due to laser aging or slow variations in temperature or bias
25 voltage) can be eliminated. Several dither cycles can be averaged together to further improve the signal quality.

First, the physical principles underlying the first and second derivative detection schemes, is now described by way of example of a length of optical fiber provided in a

DWDM optical communication system having impurities that selectively absorb light at wavelengths being used for DWDM optical communications. This absorption may be modeled as one or more “color centers” distributed along a path length x within the fiber, with each absorption center having an absorptivity $a(\lambda)$ with a concentration C of

5 absorbers in the fiber. The Beer’s Law relationship describing the exponential relation between the light intensity after passing through a length x of absorbing fiber, $I(x, \lambda)$, and the previous unattenuated intensity $I(0, \lambda)$, is expressed in equation 1) as follows:

$$I(x, \lambda) = I_0(0, \lambda) \exp(-a(\lambda) C x) \quad (1)$$

10 Now, letting the intensity be time varying in a known way to represent the transmission of data or dither modulation, as described by the modulation function $F(t)$. This variation is expressed in equation 2) as:

$$15 \quad I(x, \lambda, t) = F(t) I(x, \lambda) \quad (2)$$

A first derivative of this relation is set forth according to equation 3) as:

$$dI(\lambda)/d \lambda = F(t) \exp(-a(\lambda) C x) [dI_0(\lambda)/d \lambda - I_0(\lambda) da(\lambda)/d \lambda] \quad (3)$$

20 and the second derivative yields equation 4) as follows:

$$25 \quad d^2I(\lambda)/d \lambda^2 = F(t) \exp(-a(\lambda) C x) [d^2I_0(\lambda)/d \lambda^2 - 2Cx dI_0(\lambda)/d \lambda da(\lambda)/d \lambda + I_0(\lambda)c^2x^2 d^2a(\lambda)/d \lambda^2] \quad (4)$$

which may be rearranged according to equation 5) as follows:

$$30 \quad d^2I(\lambda)/I(\lambda)d \lambda^2 = (1/I_0(\lambda)) [d^2I_0(\lambda)/d \lambda^2 - Cx [I_0(\lambda) d^2a(\lambda)/d \lambda^2 + 2dI_0(\lambda)/d \lambda da(\lambda)/d \lambda] + c^2x^2 I_0(\lambda)da(\lambda)/d \lambda] \quad (5)$$

According to the invention, the first and second derivatives may be measured by selection of a lock-in amplifier peak to the dither frequency or twice the dither frequency, 35 respectively. For most practical absorbers, particularly with sufficiently small dither intervals, $a(\lambda)$ is a slowly varying function of wavelength. If the $a(\lambda)$ term is

approximated as a constant, then the third term (i.e., $(1/I_o(\lambda)) c^2 x^2 I_o(\lambda) da(\lambda)/d \lambda$) in equation (5) above may be neglected compared with the second term; thus, the result of a single dither scan yields a signal proportional to Cx ; since x is a known quantity, the absorption concentration C may be determined and used to compute the optical attenuation vs. 5 wavelength. It is understood that this calculation may be performed by a digital circuit in real time if required for the application.

It should be understood that the same approach may be used to determine the accuracy 10 with which the wavelength of interest is aligned with any attenuating mechanism in its path, including a wavelength selective filter in a DWDM receiver.

The apparatus for implementing derivative techniques in DWDM systems according to the principles of the invention is designed to dither either the location of an optical filter on a mechanical mount, such as shown in Figure 1(a), or to dither the optical wavelength 15 with respect to the filter bandpass center frequency such as shown in Figure 1(b). As shown in Figure 1(a), the derivative measurement principle according to a first embodiment is applied to an optical system 100 comprising a tunable bandpass filter or wavelength selective device 250 mounted on a servo and having a variable or modulated peaked passband function. It should be understood that the filter may be optionally 20 connected to a fiber link (not shown) and further may be integrated as part of a receiver device. The filter may comprise an electro-optic filter, diffraction grating, prism, fiber Bragg grating, integrated optics interferometer, electroabsorption filter and the like. Specifically, as shown in Figure 1(a), the system 100 includes an optical signal generator, e.g., VCSEL laser diode device 110, for providing laser light 160 having a fixed center 25 wavelength for transmission via an optical communication channel formed in the system. The laser diode 110 generates an optical signal 160 in response to application of a bias voltage 142 provided by a bias voltage generator device 140. As further shown in Figure 1(a), there is provided an external oscillator 220 which generates a low frequency dither modulation signal 270 (e.g., a sinusoidal signal in the kHz frequency range or less,

however adjustable as needed, in order to avoid interference with other operating frequencies in the system). As described in greater detail herein, the low frequency dither modulation signal 270 is imposed on the tunable filter device to modulate the filter response at a dither modulation frequency about its peaked passband center frequency.

- 5 According to the WLL principle, the optical signal out of diode 110 is input to the dither modulated filter device 250. As further shown in Figure 1(a), the resulting dither-modulated filtered light signal 290 output from the tunable optical filter 250 is input to a commercially available photodetector device such as P-i-N diode 300 for detecting the level of optical signal light. As further shown in Figure 1(a), the photodetector device
10 300, in response, generates an electric signal 320 that is proportional to the amount of light detected. The electrical signal 320 is amplified by lock-in amplifier device 350 at a frequency corresponding to either the dither frequency, when the system is implemented for measuring the first derivative of the light intensity versus wavelength, or two times the dither frequency for measuring the second derivative of the light intensity versus
15 wavelength. After locking in to the dither or its harmonic, the resulting signal 370, i.e., $I(x, \lambda)$, signal is fed back to a mixer device 400, where it is multiplied with the low frequency modulation signal 270 from the same external oscillator 220 used to dither modulate the filter frequency response at the low modulation frequency. This results in the cross product of the two signals as described herein. Implementing low pass filter
20 450, the resulting signal is low pass filtered to remove higher order terms, and then integrated and optionally digitized by device 480 to result in an error signal 500 which is representative of the first derivative of the optical signal at the fixed wavelength and which may be input to a control circuit 512 for determining the amount of attenuation of the optical system at the fixed laser wavelength. According to the error signal 500
25 detected, the peaked passband function of the filter may be accordingly adjusted until the error signal is zero which indicates a maximum power transfer matching of the laser signal center frequency with the tuned filter passband center frequency. It should be understood that the second derivative measurement may also be used however, instead of

locking in on the dither frequency, a harmonic of the dither frequency, i.e., at two times dither, may be used to adjust the tunable filter.

In the second embodiment, depicted in Figure 1(b), the tunable filter is replaced by a
5 frequency selective device 250' such as a bandpass filter having a peaked passband function or may comprise a fiber with color centers, or the like. Thus, in the second embodiment, rather than adjusting a tunable filter, the first derivative calculation is used to modify the bias signal 142 applied to the laser diode 110 in order to adjust the center frequency of the source signal. That is, as shown in Figure 1(b), the external oscillator
10 220 generates a low frequency dither modulation signal 270 (e.g., a sinusoidal signal in the kHz frequency range or less, however adjustable as needed, in order to avoid interference with other operating frequencies in the system) that is imposed on the bias signal generator 140 in order to dither modulate the optical signal 160 about its peaked center wavelength. The laser voltage bias is thus modulated by a dithering current from
15 the signal generator. Dither modulation of the bias voltage generator produces a corresponding dither in the light output of the filter 250' (of fixed passband response) which is monitored by the photodetector 300. In response, the photodetector generates the dither-modulated electrical signal 320 that is proportional to the amount of light detected. The electrical signal 320 is amplified by lock-in amplifier device 350 at a
20 frequency corresponding to either the dither frequency, when the system is implemented for measuring the first derivative of the light intensity versus wavelength, or two times the dither frequency for measuring the second derivative of the light intensity versus wavelength. After locking in to the dither or its harmonic, the resulting signal 370, i.e., $I(x, \lambda)$, signal is fed back to the mixer device 400, where it is multiplied with the low
25 frequency modulation signal 270 from the same external oscillator 220 used to dither modulate the laser diode at the low modulation frequency. This results in the cross product of the two signals as described herein. Implementing low pass filter 450, the resulting signal is low pass filtered to remove higher order terms, and then integrated and optionally digitized by device 480 to result in an error signal 500 representative of the

first derivative of the optical signal at the adjusted wavelength which may be input to a control circuit 512 for determining the amount of attenuation of the optical system at the adjusted laser wavelength. According to the error signal 500 detected, the center frequency of the laser signal may be accordingly adjusted until the error signal is zero
5 which indicates a maximum power transfer matching of the adjusted laser signal center frequency with the filter fixed passband center frequency. It should be understood that the second derivative measurement may also be used however, instead of locking in on the dither frequency, a harmonic of the dither frequency, i.e., at two times dither, may be used to adjust the tunable filter.

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It should be understood that, in both embodiments depicted in Figures 1(a) and 1(b), the optical dither feedback loop may be used to measure the first or second derivatives of the light intensity vs. wavelength functions given in the equations 3) and 5) herein. If there is any change in $I(x, \lambda)$ over time (for example, as a result of adding or dropping
15 wavelengths on the network or changing optical filters with different transfer functions) the change occurs more quickly in the higher order derivatives of $I(\lambda)$. Thus, it may also be compensated more quickly by making changes in the variables which affect equation (5) above.

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It should be understood that while the approach of the invention may be used to determine the accuracy with which the wavelength of interest is aligned with any attenuating element in its path in a DWDM receiver, the principles of the invention may be applied for applications including: 1) WDM interleavers and 2) dynamic background subtraction for optical amplifiers with adaptive gain and channel equalizers. Thus, by
25 introducing wavelength sweeping or tuning, dithering may be applied at desired wavelength values in a quasi-continuous manner, thus extending applicability of the invention from discrete spectral lines to multi-line spectra.

While the invention has been particularly shown and described with respect to illustrative and preformed embodiments thereof, it will be understood by those skilled in the art that the foregoing and other changes in form and details may be made therein without departing from the spirit and scope of the invention which should be limited only by the
5 scope of the appended claims.